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Method and transmitter for transmitting data in a multicarrier system via a number of transmitting antennas

The invention relates to a method for transmitting data by radio in accordance with the preamble of claim 1 and of claim 3.

The invention further relates to a transmitter for transmitting data by radio via a number of antennas in accordance with the preamble of Claim 8.

In radio communication systems data (for example speech, picture information, video information, SMS (Short Message Service) or other data) is transmitted with the aid of electromagnetic waves via a radio interface between sending and receiving station. The electromagnetic waves in such systems are radiated using carrier frequencies which lie within the frequency range provided for the relevant system. A radio communication system in this case includes subscriber stations, e.g. mobile stations, base stations, e.g. node Bs or other radio access devices, as well as further network-side devices where required.

To guarantee that data is transmitted as efficiently as possible the overall frequency band available is divided up into a number of subcarriers (multi-carrier method). The basic idea underlying multi-carrier systems, also referred to as OFDM (Orthogonal Frequency Division Multiplexer) systems, is to translate the initial situation of the transmission of a broadband signal into the transmission of a quantify of narrowband orthogonal signals.

With OFDM time pulse shapes which are almost rectangular are used for the subcarriers. The frequency spacing of the subcarriers is selected such that in the frequency space for that frequency at which the signal of a subcarrier is evaluated, the signals of the other subcarriers exhibit a zero crossing. The subcarriers are thus orthogonal to each other. A spectral overlapping of the subcarriers and as a result a high packing density of the subcarriers is allowed, since the orthogonality ensures that the individual subcarriers can be distinguished. A better spectral efficiency than with simple FDM (Frequency Division Multiplexing) is thus achieved."

Because of the dispersion behavior of radio channels it is advantageous, in OFDM systems, to introduce a diversity, into the space area for example, in order to increase the reliability of data transmission. To do this a plurality of antennas are used on the transmit side which each transmit the same information. On receipt the signals subject to different fluctuations are combined. Examples of such methods are the CDD (Cyclic Delay Diversity) method presented in

A. Damman, S. Kaiser: Low Complex Standard Conformable
Antenna Diversity Techniques for OFDM Systems and its
Application to the DVB-T System, Proceedings of the 4 th
International ITG Conference on Source and Channel Coding,
Berlin 2002,

in which virtual echoes of the signal sent on an antenna are created by the different antennas. However in

A. Huebner, M. Bossert, F. Schuehlein, H. Haas, E. Costa: On Cyclic Delay Diversity in OFDM Based Transmission Schemes, Proceedings of the 7th International OFDM Workshop, Hamburg 2002

it is shown that this method can only be viably used in combination with channel encoding, such as convolutional encoding, to reduce transmission errors.

Furthermore, for OFDM transmission with a plurality of antennas the Alamout method in accordance with

S.M. Alamouti: A .Simple Transmit Diversity Technique for wireless Communications, IEEE Journal an Selected Areas in Communications, vol. 16, no. 8, P. 1451-1458, October 1998 can be used, in which two consecutive symbols are processed on the transmit side such that two transmit antennas transmit signals which are orthogonal to each other. A disadvantage of the Alamouti method has proved to be the need to use an expensive, modified demodulator on the receive side, and that a plurality of antennas cannot be used for the transmission without resulting in restrictions as regards the choice of the modulation method.

The object of the invention is to demonstrate a method and a transmitter of the type specified at the start which allows an efficient transmission of data in a multi-carrier system while utilizing the space diversity.

This object is achieved as regards the method by a method with the features of claim 1.

Advantageous embodiments and further developments are the object of the subclaims.

In the inventive method for transmitting data by radio a frequency band divided into a plurality of subcarriers and a plurality of antennas is used for transmission. The data is divided up in accordance with the number of the plurality of subcarriers into a number of elements to be transmitted by each antenna in each case. In this case each element is assigned a subcarrier in ecah case for transmission for each antenna. At least two antennas transmit different elements on at least one subcarrier. In accordance with the invention, before an OFDM modulation, for each antenna each element is multiplied by an antenna-specific and element-specific factor.

The number of elements of the data corresponds to the number of subcarriers. Each antenna transmits each element on a subcarrier, with not all antennas transmitting the elements on the same subcarriers. In particular it is possible for an element to be transmitted from each of the antennas on a different subcarrier. This can also apply to all elements. The OFDM modulation serves to prepare the signal for transmission, it comprises an inverse Fourier transformation in which the signal is transformed from the frequency domain into the time domain. Before the OFDM modulation each element to be transmitted by an antenna is multiplied by an antenna-specific and element-specific factor. This factor differs as a rule for all elements to be transmitted by an antenna and also for each element to be transmitted by different antennas. It is however also possible for it to be the same for different elements to be transmitted by one or more antennas, in which case however the relevant element and the relevant antenna must basically be included in the specifications in which the factor is defined.

The factor is in particular a complex or real number amounting to 1. This can be implemented by multiplication with an exponential function with a corresponding complex or real component. This multiplication corresponds to a phase shift of a signal.

The above object as regards the method is further achieved by a method with the features of claim 3.

Advantageous embodiments and further developments are the object of the subclaims.

In the method for transmitting data by radio a frequency band divided up into a plurality of subcarriers and a plurality of antennas is used for transmission. The data is divided up in

accordance with the number of the plurality of subcarriers into a number of elements to be transmitted by each antenna in each case. For each antenna each element is assigned to a subcarrier for transmission in each case. At least two antennas transmit different elements on at least one subcarrier. In accordance with the invention, after an OFDM modulation, there is a re-arrangement of the time sequence for at least one antenna of the time-dependent signals produced because of the OFDM modulation.

Whereas in the method described first the multiplication by the antenna-specific and element-specific factor is undertaken before the OFDM modulation, the re-arrangement of the time sequence is undertaken in the method described as the second method after the OFDM modulation. These two solutions for achieving the above object are however equally valid. Thus it can be shown mathematically that the re-arrangement of the timing sequence of the signal after the OFDM modulation is of equal significance to the multiplication of the signal before the OFDM modulation by a corresponding antenna-specific or element-specific factor.

In particular the re-arrangement of the timing sequence occurs for at least two antennas in accordance with a common pattern. It is especially advantageous for the re-arrangement of the timing sequence to be undertaken in accordance with a common pattern for all antennas. A pattern in this case means a set of specifications in accordance with which the re-arrangement is performed. The common pattern can for example involve a cyclic permutation. With a cyclic permutation, parts of the signal are interchanged in their sequence in a regular way such that once a cycle is completed in the given case after the number of permutations which corresponds to the number of the antennas, the original sequence is restored.

In an embodiment of the invention the assignment of the elements to subcarriers is undertaken for at least two antennas in accordance with a common pattern, such as for example in accordance with a cyclic permutation. it is especially advantageous if for all antennas the assignment of the elements to subcarriers is undertaken according to a common pattern.

The method in accordance with the invention leads to a marked improvement compared to the prior art described above.

The above object as regards the transmit device is achieved by a transmit device with the features of claim 8.

The transmit device for transmitting data by radio via a number of antennas uses a frequency band subdivided into a plurality of subcarriers for transmission. The transmit device includes means for dividing up the data into a number of elements corresponding to the number of the plurality of subcarriers to be transmitted by each antenna in each case, as well as means for assigning the elements to a subcarrier in each case for transmission for each antenna such that at least two antennas transmit different elements on at least one subcarrier.

In accordance with the invention the transmit device further includes means for multiplying each element for each antenna by an antenna-specific and element-specific factor before OFDM modulation. Alternatively it can include means for rearranging the timing sequence of the time-dependent signals produced as a result of the OFDM modulation for at least one antenna after the OFDM modulation.

The transmit device in accordance with the invention is especially suited to carrying out the inventive method. It can

feature further means for doing so.

The invention is explained in more detail below on the basis of an exemplary embodiment . The Figures show

Figure 1: a schematic diagram of the sequence of the inventive method,

Figure 2a: a first inventive transmitter,

Figure 2b: a second inventive transmitter.

The exemplary embodiment relates to an OFDM system, e.g. in accordance with the standards IEEE 802.16a or HIPERLAN/2. The transmission of data using three transmit antennas is considered.

In Figure 1, from the data S to be sent a vector with elements S_1 , S_2 and S_3 is formed. The elements S_1 , S_2 and S_3 are symbols which are each to be transmitted on a subcarrier of the frequency bandwidth available for transmitting the data. In this case a larger number of subcarriers than three can be available in the OFDM system for data transmission, but these are not relevant to the invention.

The vector with elements S_1 , S_2 and S_3 is fed into three branches which come out into three antennas TX1, TX2 and TX3.. In a first step the vector with the elements S_1 , S_2 and S_3 is subjected to a cyclic shift or permutation in accordance with pattern a, with the shift within the first branch representing an identity operation and thus is not shown in Figure 1, while the shift in the second and the third branch is undertaken in accordance with specifications a_2 and a_3 . The result of this first step is then the following matrix:

$$\vec{S}_{a} = \begin{pmatrix} S_{1} & S_{2} & S_{3} \\ S_{2} & S_{3} & S_{1} \\ S_{3} & S_{1} & S_{2} \end{pmatrix}$$

The first column of the matrix S_a in this case represents the symbols (S_1,S_2,S_3) which are to be transmitted via the first antenna TX1, the second and the third columns accordingly contain those symbols (S_2,S_3,S_1) and (S_3,S_1,S_2) which are to be transmitted by the second and the third antennas TX2 and TX3. The first row of the matrix S_a contains those symbols (S_1,S_2,S_3) which are to be transmitted on the first subcarrier, in the second and third row are those symbols (S_2,S_3,S_1) and (S_3,S_1,S_2) which are to be transmitted on the second and the third subcarrier It can be seen that each antenna transmits each element S_1 , S_2 and S_3 on precisely one subcarrier, with none of the elements S_1 , S_2 and S_3 of a number of antennas being transmitted on the same subcarrier.

The common pattern a underlying the assignment of the elements S_1 , S_2 and S_3 to the subcarriers of the cyclic shift in a frequency space for the three antennas TX1, TX2 and TX3 operates such that for the first antenna TX1 the three elements S_1 , S_2 and S_3 are distributed to the subcarriers in their original sequence. For the second antenna TX2 the first element S_1 occupies the last place in the above matrix diagram a, with the second and the third element S_2 and S_3 moving forward to the first two subcarriers, so that overall compared to the assignment of first antenna TX1 each element S_1 , S_2 and S_3 has been shifted upwards by one subcarrier. For the third antenna TX3 a further shift of the sequence of the second antenna TX2 upwards by one subcarrier occurs in the above matrix diagram.

Subsequently the OFDM modulation OFDM occurs in Figure 1. In

this case the frequency-dependent signals of the individual antenna branch each undergo an inverse Fourier transformation and a parallel-serial conversion.

The result of this is a time-dependent signal. In the form of a matrix this signal can be represented as follows:

$$\vec{S}_{OFDM} = egin{pmatrix} q_1 & q_4 & q_7 \\ q_2 & q_5 & q_8 \\ q_3 & q_6 & q_9 \end{pmatrix}$$

The first column of the matrix \hat{S}_{OFDM} after OFDM modulation OFDM has been completed In this case represents the symbols (q_1,q_2,q_3) which are to be transmitted via the first antenna TX1 the second and the third columns accordingly contain the symbols (q_4,q_5,q_6) and (q_7,q_8,q_9) which are to be transmitted by the second and the third antenna TX2 and TX3. In the first row of the matrix \hat{S}_{OFDM} there are those symbols (q_1,q_4,q_7) which are to be transmitted at a first point in time, in the second and third rows those symbols (q_2,q_5,q_8) and (q_3,q_6,q_9) which are to be transmitted at a second and third point in time.

In a further processing step the matrix \hat{S}_{OFDM} is processed in accordance with pattern b which in its turn corresponds to a cyclic shift. The shift within the first branch represents an identification operation and is thus not shown in Figure 1, while the shift in the second and the third branch is undertaken in accordance with specifications b_2 and b_3 . The cyclic shift is performed in accordance with the sequence described above, resulting in the following matrix:

$$\vec{S}_b = \begin{pmatrix} q_1 & q_5 & q_9 \\ q_2 & q_6 & q_7 \\ q_3 & q_4 & q_8 \end{pmatrix}.$$

Before the data S is transmitted in the form of the matrix \hat{S}_b a guard period is inserted for each antenna TX1, TX2 and TX3 to reduce interference to the subsequent data to be transmitted. The antennas TX1, TX2 and TX3 then transmit the symbols in the corresponding sequence, for the first antenna TX1 that is q_1 for example, followed by q_2 , q_3 and the guard period.

The method in accordance with the invention has thus combined both a cyclic shift in the frequency range, i.e. over the different subcarriers, with of a cyclic shift in the time range.

The second step of the cyclic shift in time range according to pattern b can also be achieved other than in the way shown in Figure 1 by multiplying the individual entries $S_{a,kl}$ of the matrix S_a by a particular factor in accordance with

$$\bar{S}_{b,kl}^{beforeOFDM} = S_{a,kl} \cdot e^{-i\frac{2\pi}{N}l\delta_k}$$

Subsequently each entry $S_{a,kl}$ of the matrix S_a , indexed by k and l is multiplied by an exponential function. The variable N in the exponent stands for the number of antennas in this case. The whole number shift \ddot{a}_k is subject to the condition

 $0 \le \ddot{a}_k \le N-1$.

For two antennas the shift \ddot{a}_k can amount to e.g. 0 and $\frac{N}{2}$.

The exponential function used as a factor is element-specific and antenna-specific since the two indices k and l are components of the exponent. The shiftä $_k$ can be defined so that in its final effect it corresponds to the pattern b. Then the result of the inverse Fourier transformation on OFDM

modulation from the matrix $\vec{S}_b^{\ beforeOFDM}$ is the matrix shown above \vec{S}_b .

Figures 2a and 2b above each show an inventive transmitter S with three antennas TX1, TX2 and TX3 . This comprises means $\mbox{M1}$ for dividing the data into elements and means M2 for assigning the elements to a subcarrier for each antenna. The assignment is undertaken here in accordance with the cyclic shift described above. Figure 2a shows the case where the second processing step occurs before the OFDM modulation. To this end the transmitter S features means M3 for multiplying the elements for each antenna by the antenna-specific and elementspecific exponential factor. By contrast the second processing step occurs in the transmitter S of Figure 2b, as also shown in Figure 1, after the OFDM modulation. The transmitter S of Figure 2b includes for this purpose means M4 to rearrange the timing sequence of the signal obtained by the OFDM modulation, with this rearrangement corresponding to the cyclic shift described above. To perform the OFDM modulation the two transmitters feature additional means not shown.

Although in the example described the shift of the signals in the frequency and in the time space occured in the form of a cyclic shift in each case, other patterns such as for example statistically random patterns or other shift specifications can be used to perform this action .

The signal vector R received in the receiver is produced from $R = H \cdot S + N \; , \label{eq:R}$

with H representing the matrix of the transmission scheme shown in Figure 1 which is multiplied by the data S and N representing a noise vector.

If more than two transmit antennas are used, the method in accordance with the invention reduces the bit error rates (BER) of the data transmission compared to the Alamouti method. A further advantage of the method in accordance with the invention lies in the fact that it can be used for any number of antennas. It is also possible to use any modulation alphabet in combination with the inventive method. Furthermore the method in accordance with the invention can be used with a simplified structure of the receiver compared to the prior art.